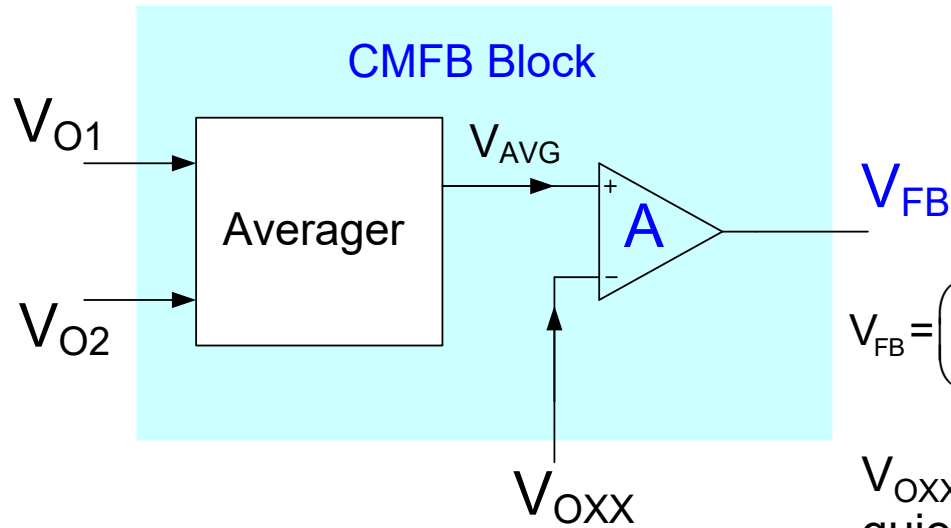
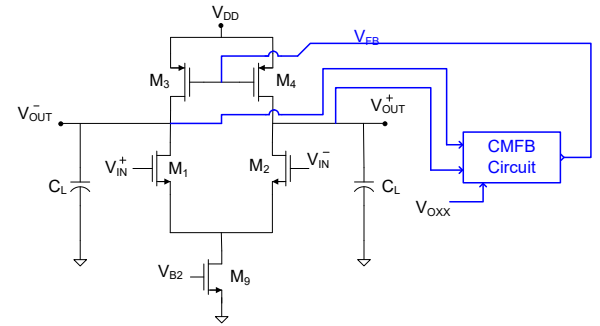
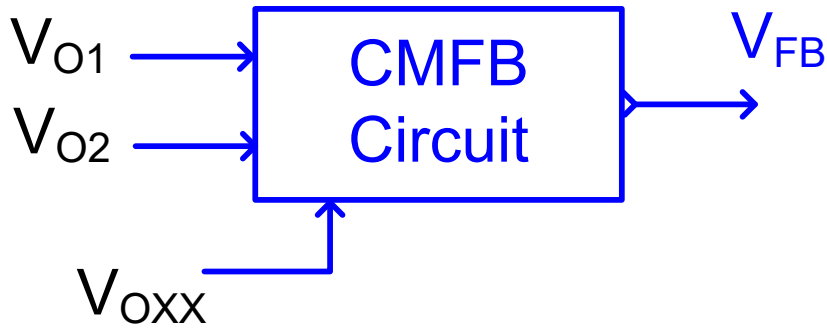


EE 435

Lecture 25

Data Converter Operation

Basic Operation of CMFB Block

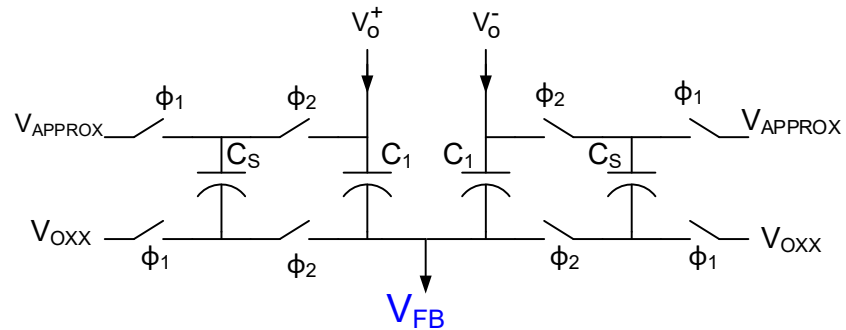
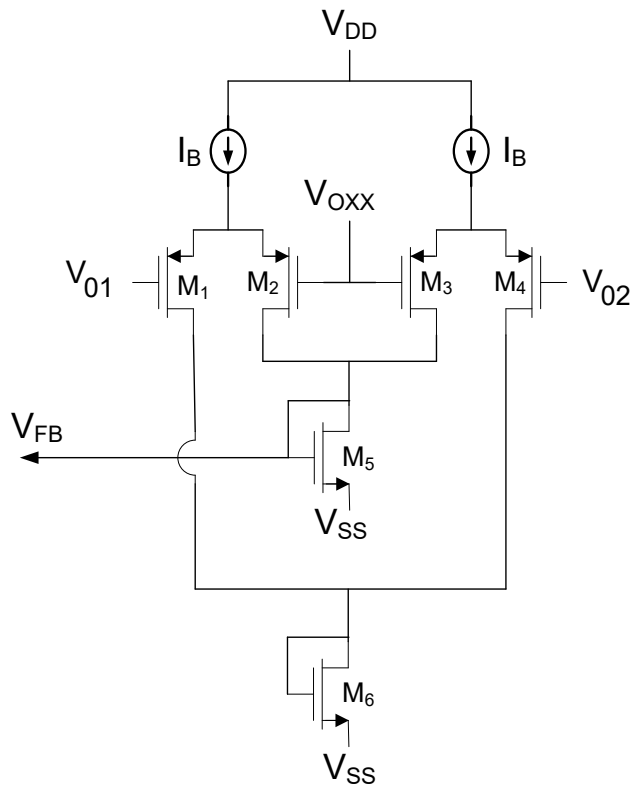
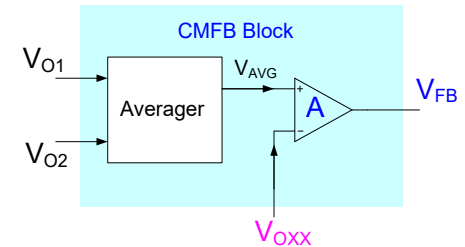


$$V_{FB} = \left(\frac{V_{O1} + V_{O2}}{2} \right) A(s)$$

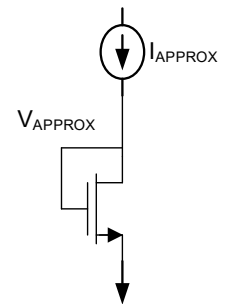
V_{OXX} is the desired quiescent voltage at the stabilization node (irrespective of where V_{FB} goes)

CMFB Circuits

- Several (but not too many) CMFB blocks are widely used
- Can be classified as either continuous-time or discrete-time



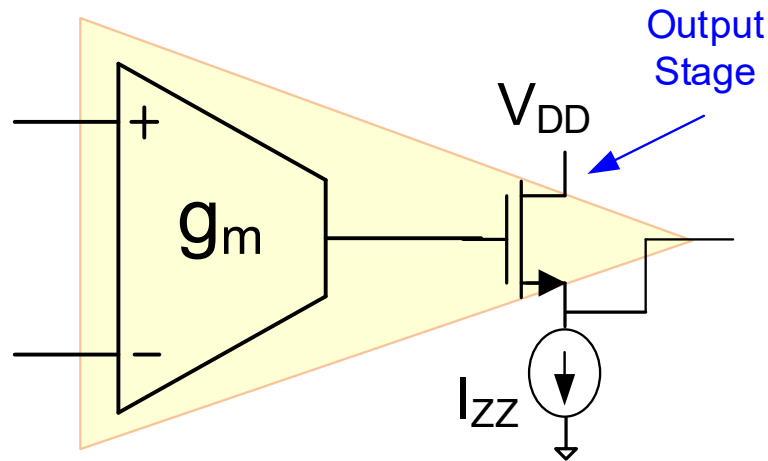
C_S small compared to C_1



- V_{OXX} generated by simple bias generator
- ϕ_1 and ϕ_2 are complimentary non-overlapping clocks that run continuously

Output Stages

The Concept:



Common Drain Amplifier Provides a Good Output Stage

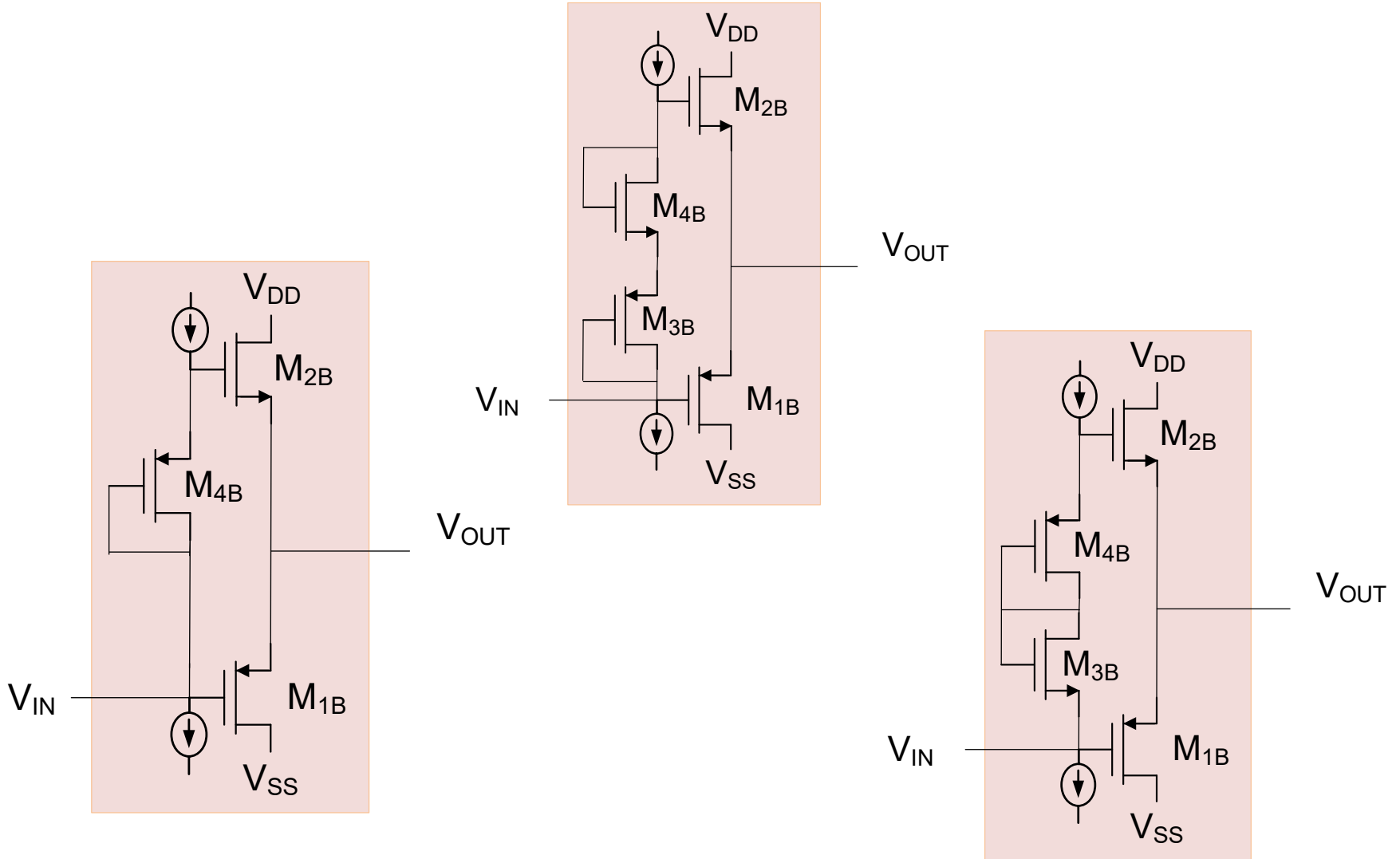
Simple structure

Some loss of output signal swing

Significant power overhead

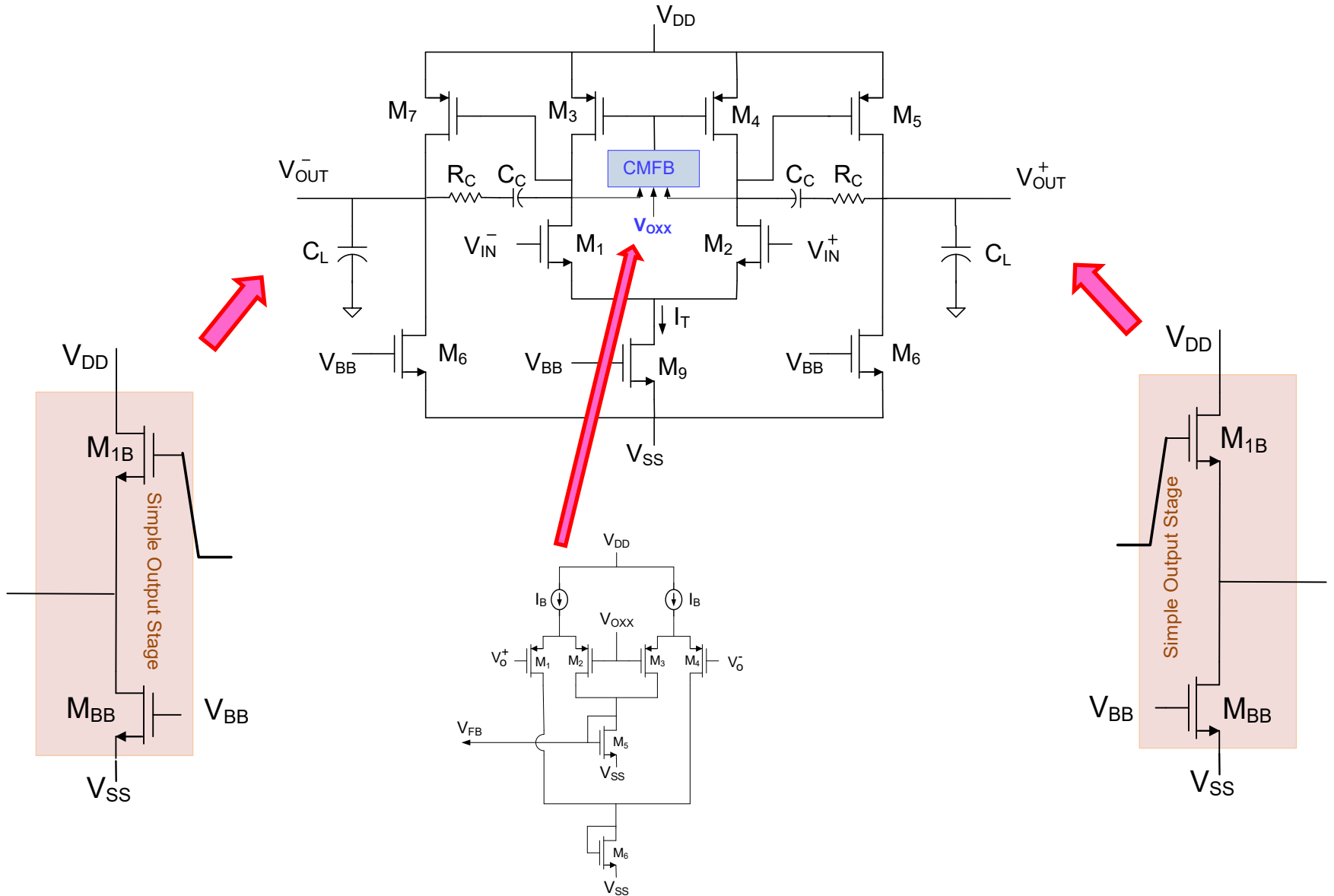
• • • • • Review from last lecture • • • • •

Other Selected Output Stages



• • • • • Review from last lecture • • • • •

Fully-Differential Op Amp with CMFB and Output Stages



Data Converters

Types:

A/D (Analog to Digital)

Converts Analog Input to a Digital Output

D/A (Digital to Analog)

Converts a Digital Input to an Analog Output

A/D is the world's most widely used mixed-signal component

D/A is often included in a FB path of an A/D

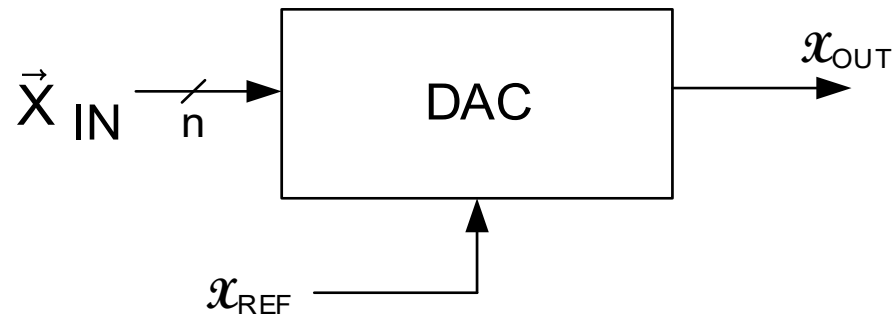
A/D and D/A fields will remain hot indefinitely

technology advances make data converter design more challenging
embedded applications

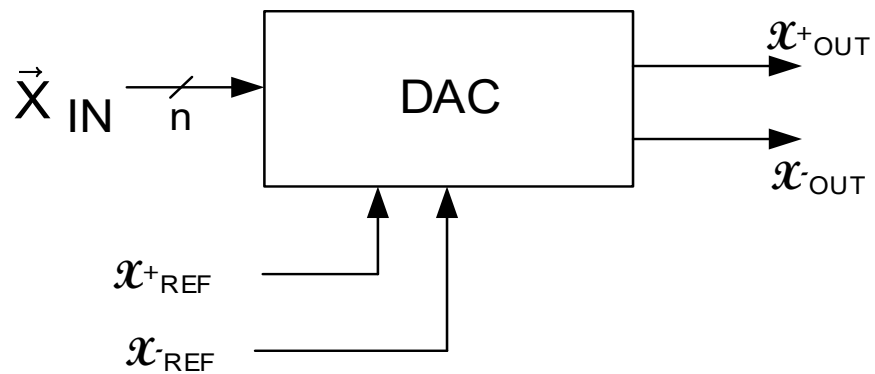
designs often very application dependent

D/A Converters

Basic structure:

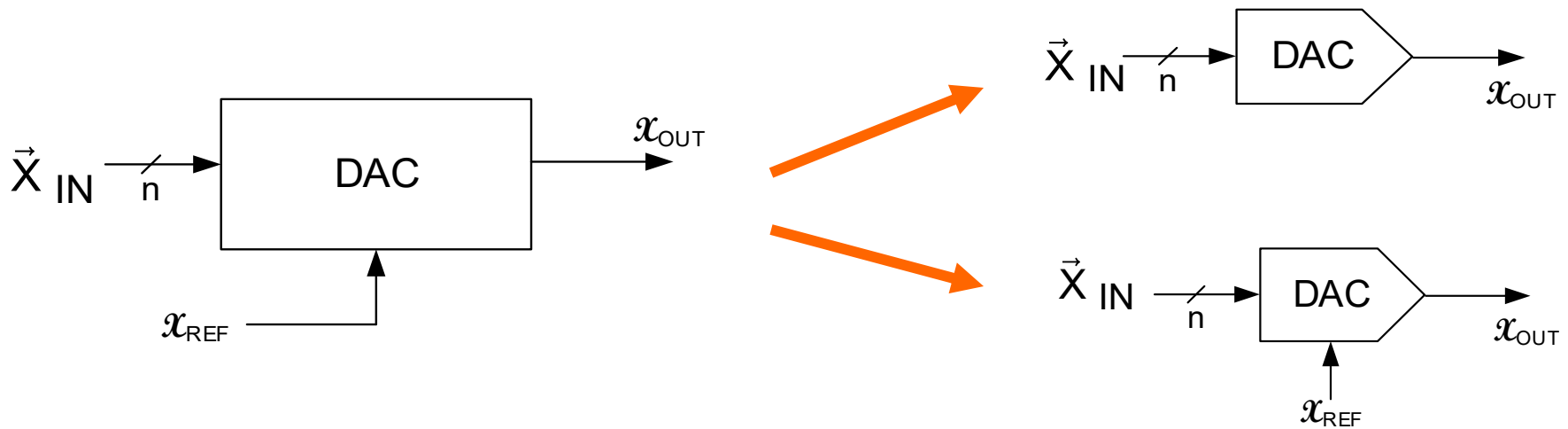


Basic structure with differential outputs::



D/A Converters

Notation:



x_{REF} is always present though often not shown on the symbol for the DAC

D/A Converters



$$\vec{X}_{IN} = \langle b_{n-1}, b_{n-1}, \dots, b_1, b_0 \rangle$$

b_0 is the Least Significant Bit (LSB)

b_{n-1} is the Most Significant Bit (MSB)

Number of ideal DAC outputs: $N=2^n$

Note: some authors use different index notation

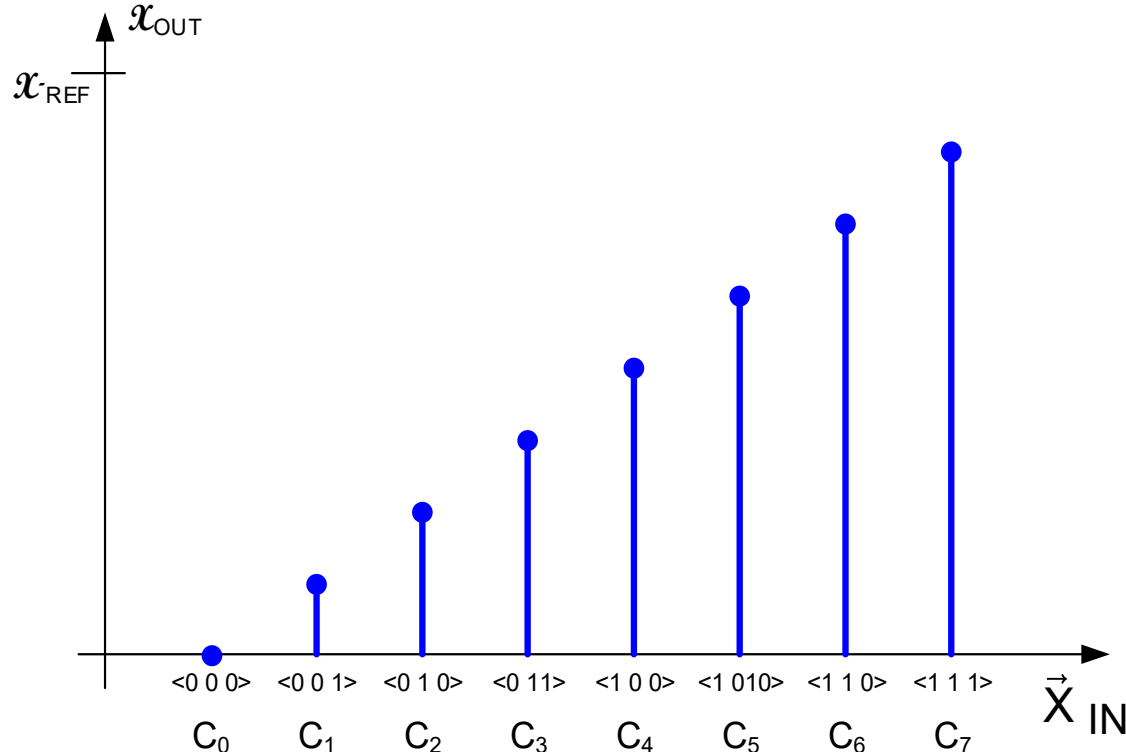
An Ideal DAC is characterized at low frequencies by its static performance

D/A Converters



$$\vec{X}_{IN} = \langle b_{n-1}, b_{n-1}, \dots, b_1, b_0 \rangle$$

An Ideal DAC transfer characteristic (3-bits)



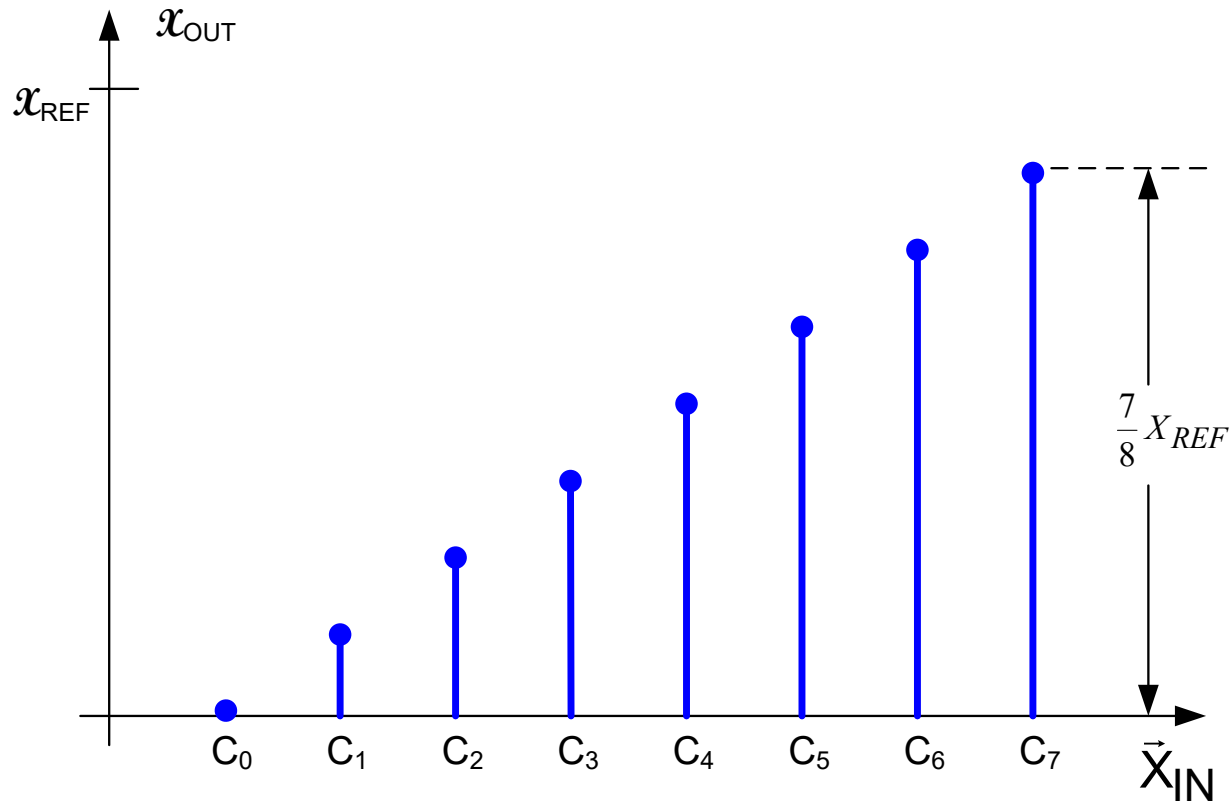
Code C_k is used to represent the decimal equivalent of the binary number $\langle b_{n-1} \dots b_0 \rangle$

D/A Converters



$$\vec{X}_{IN} = \langle b_{n-1}, b_{n-1}, \dots, b_1, b_0 \rangle$$

An Ideal DAC transfer characteristic (3-bits)

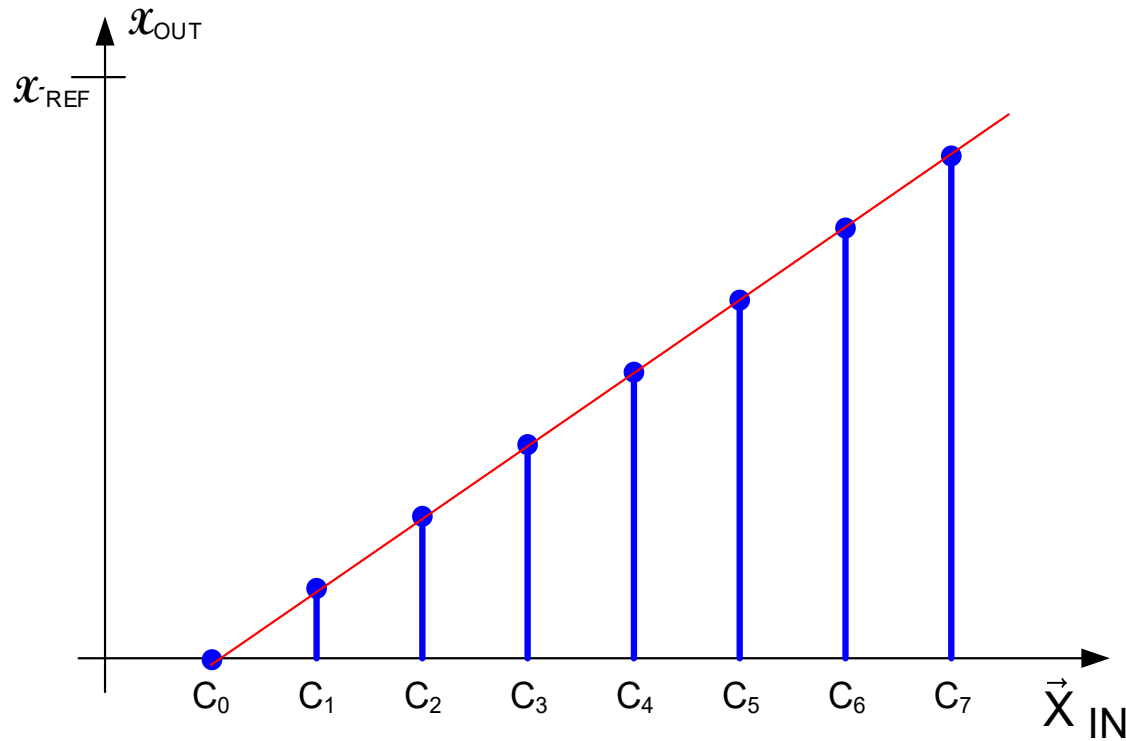


D/A Converters



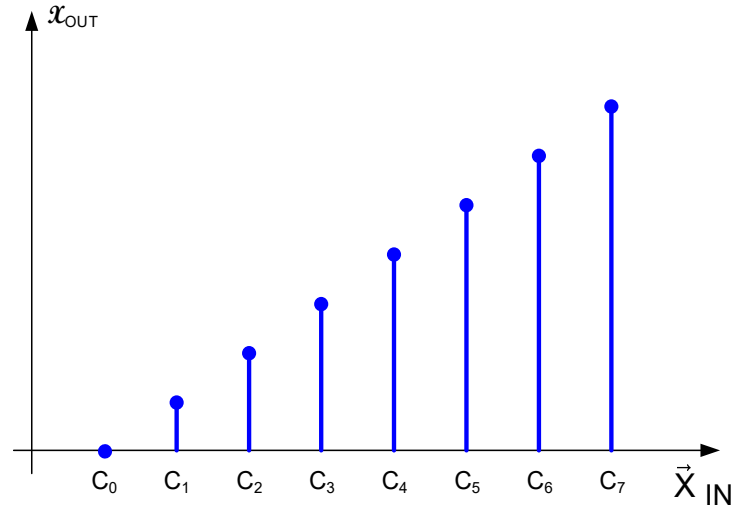
$$\vec{X}_{IN} = \langle b_{n-1}, b_{n-1}, \dots, b_1, b_0 \rangle$$

An Ideal DAC transfer characteristic (3-bits)



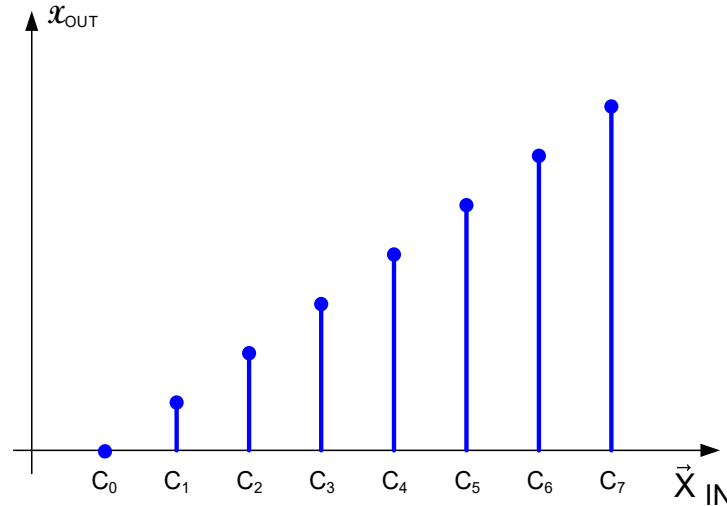
All points of this ideal DAC lie on a straight line

D/A Converters



- Most D/A ideally have a linear relationship between binary input and analog output
- Output represents a discrete set of continuous variables
- Typically this number is an integral power of 2, i.e. $N=2^n$
- \vec{X}_{IN} is always dimensionless
- x_{OUT} could have many different dimensions
- An ideal nonlinear characteristic is also possible (waveform generation and companding)
- Will assume a linear transfer characteristic is desired unless specifically stated to the contrary

D/A Converters



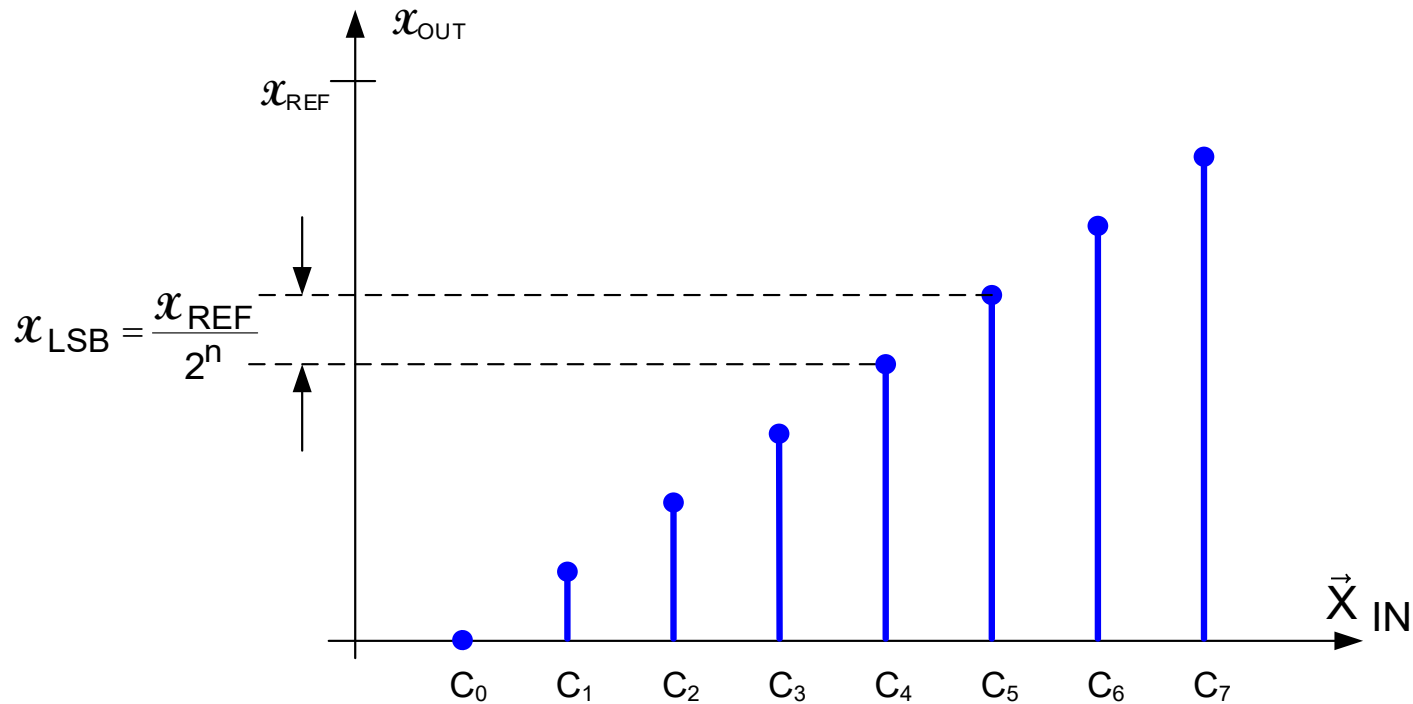
For this ideal DAC

$$x_{OUT} = X_{REF} \left(\frac{b_{n-1}}{2} + \frac{b_{n-2}}{4} + \frac{b_{n-3}}{8} + \dots + \frac{b_1}{2^{n-1}} + \frac{b_0}{2^n} \right)$$

$$x_{OUT} = X_{REF} \sum_{j=1}^n \frac{b_{n-j}}{2^j}$$

- Number of outputs gets very large for n large
- Spacing between outputs is $X_{REF}/2^n$ and gets very small for n large

D/A Converters



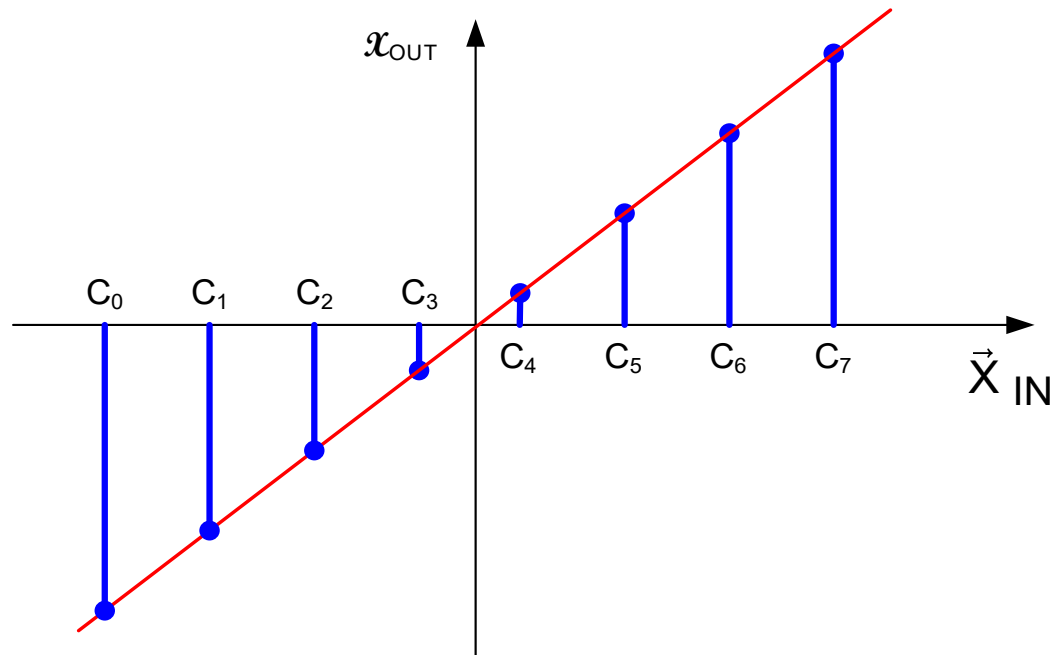
- Ideal steps all equal and termed the LSB
- x_{LSB} gets very small for small x_{REF} and large n

e.g. If $x_{REF}=1V$ and $n=16$, then $N=2^{16}=65,536$, $x_{LSB}=15.25\mu V$

D/A Converters



An alternate ideal 3-bit DAC



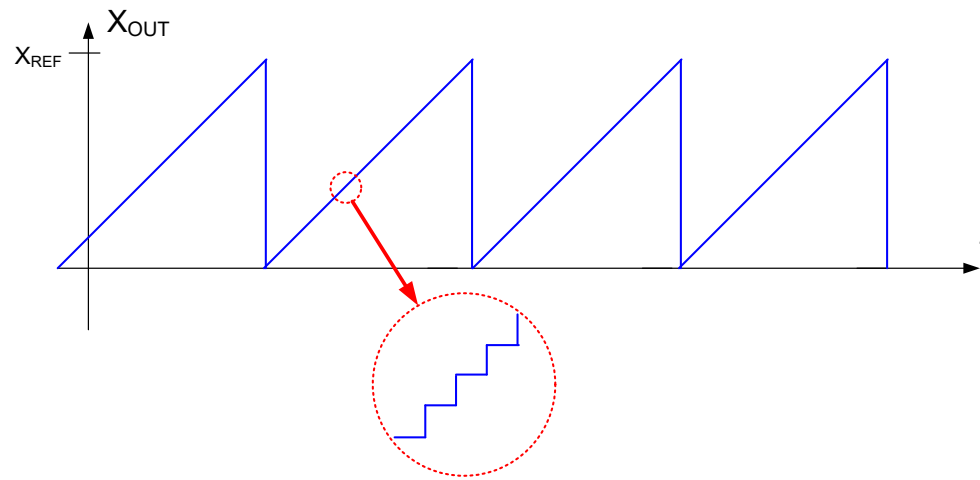
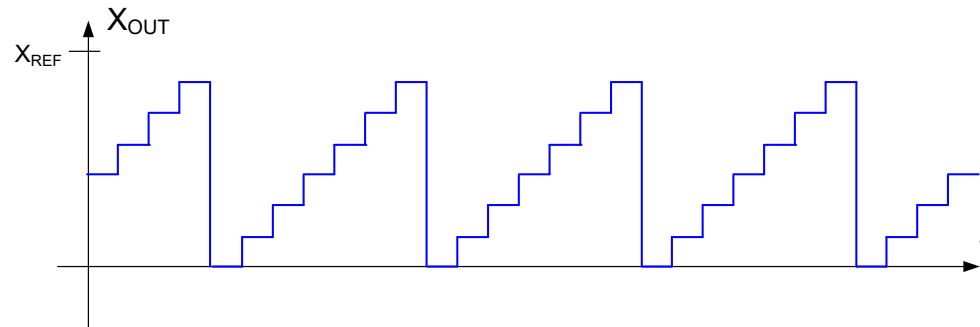
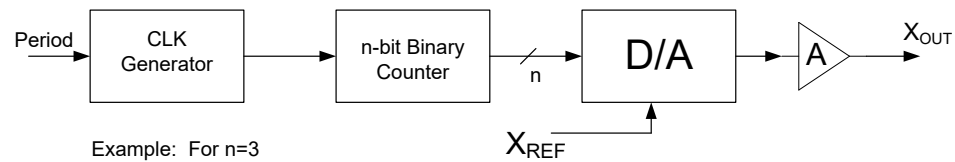
Irrespective of which form is considered, the increment in the output for one Boolean bit change in the input is x_{LSB} and the total range is 1LSB less than x_{REF}

Applications of DACs

- Waveform Generation
- Voltage Generation
- Analog Trim or Calibration
- Industrial Control Systems
- Feedback Element in ADCs
-

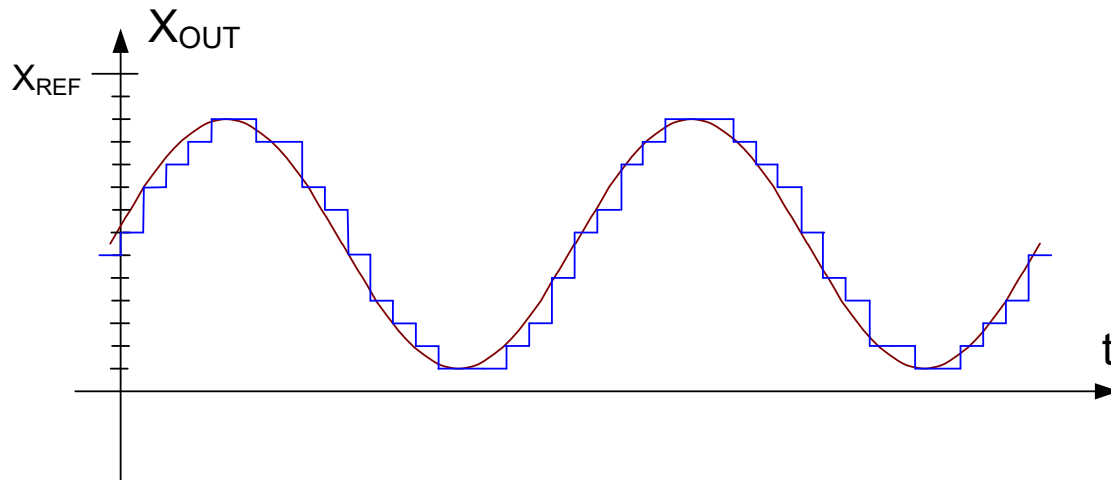
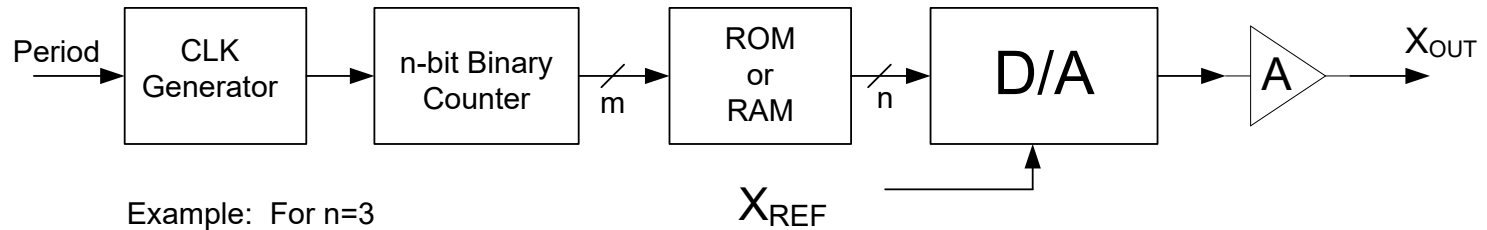
Waveform Generation with DACs

Ramp (Saw-tooth) Generator



Waveform Generation with DACs

Sine Wave Generator

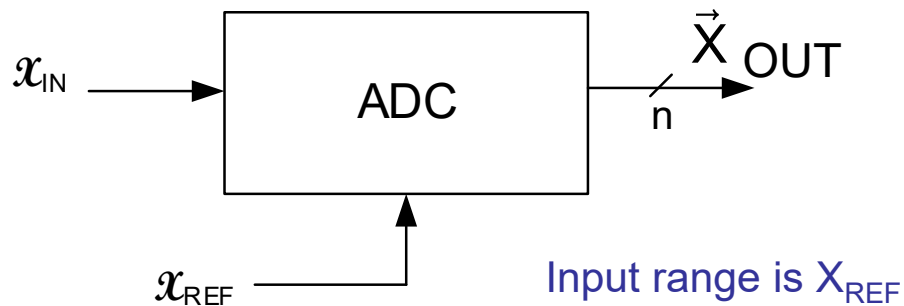


Distortion of the desired waveforms occurs due to both time and amplitude quantization

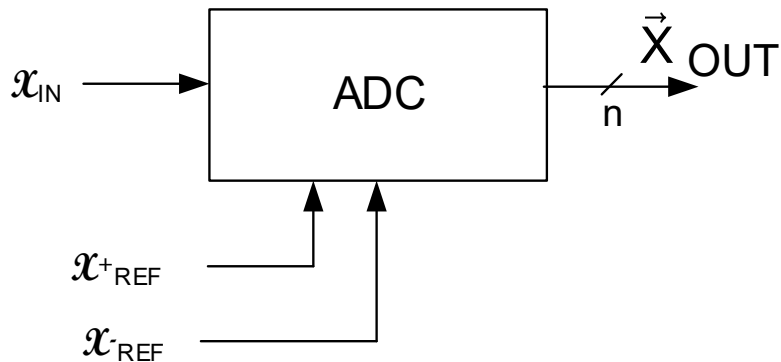
Often a filter precedes or follows the buffer amplifier to smooth the output waveform

A/D Converters

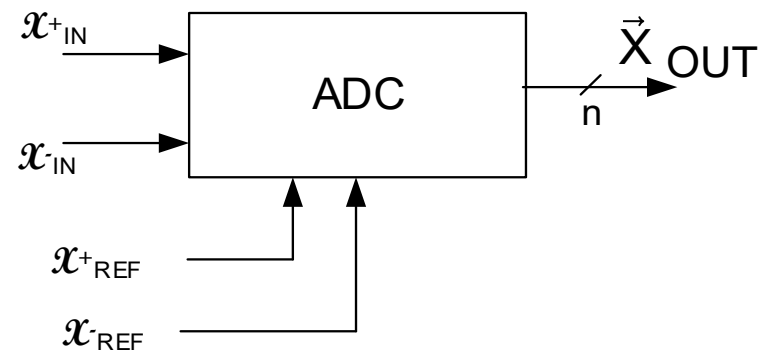
Basic structure:



Basic structure with differential inputs/references:



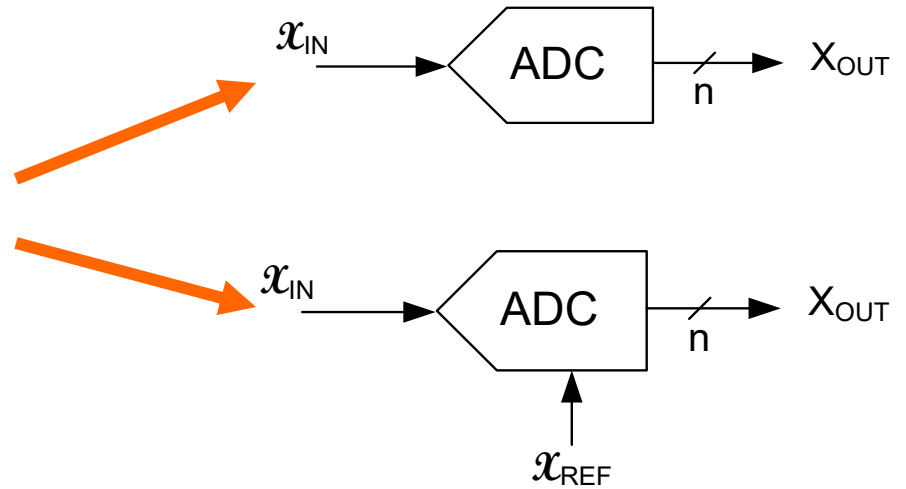
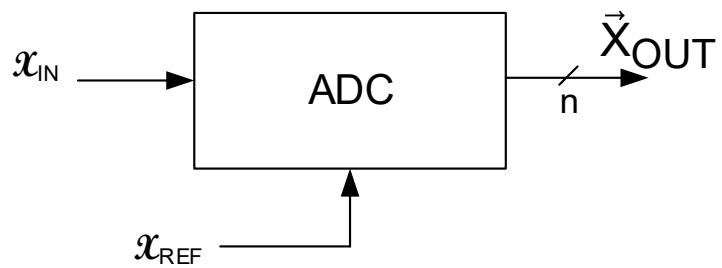
Input range is $X^+_{REF} - X^-_{REF}$



Input range is $2(X^+_{REF} - X^-_{REF})$

A/D Converters

Notation:



A/D Converters



$$\vec{X}_{OUT} = \langle d_{n-1}, d_{n-2}, \dots, d_0 \rangle$$

d_0 is the Least Significant Bit (LSB)

d_{n-1} is the Most Significant Bit (MSB)

Number of ideal ADC outputs: $N=2^n$

Note: some authors use different index notation

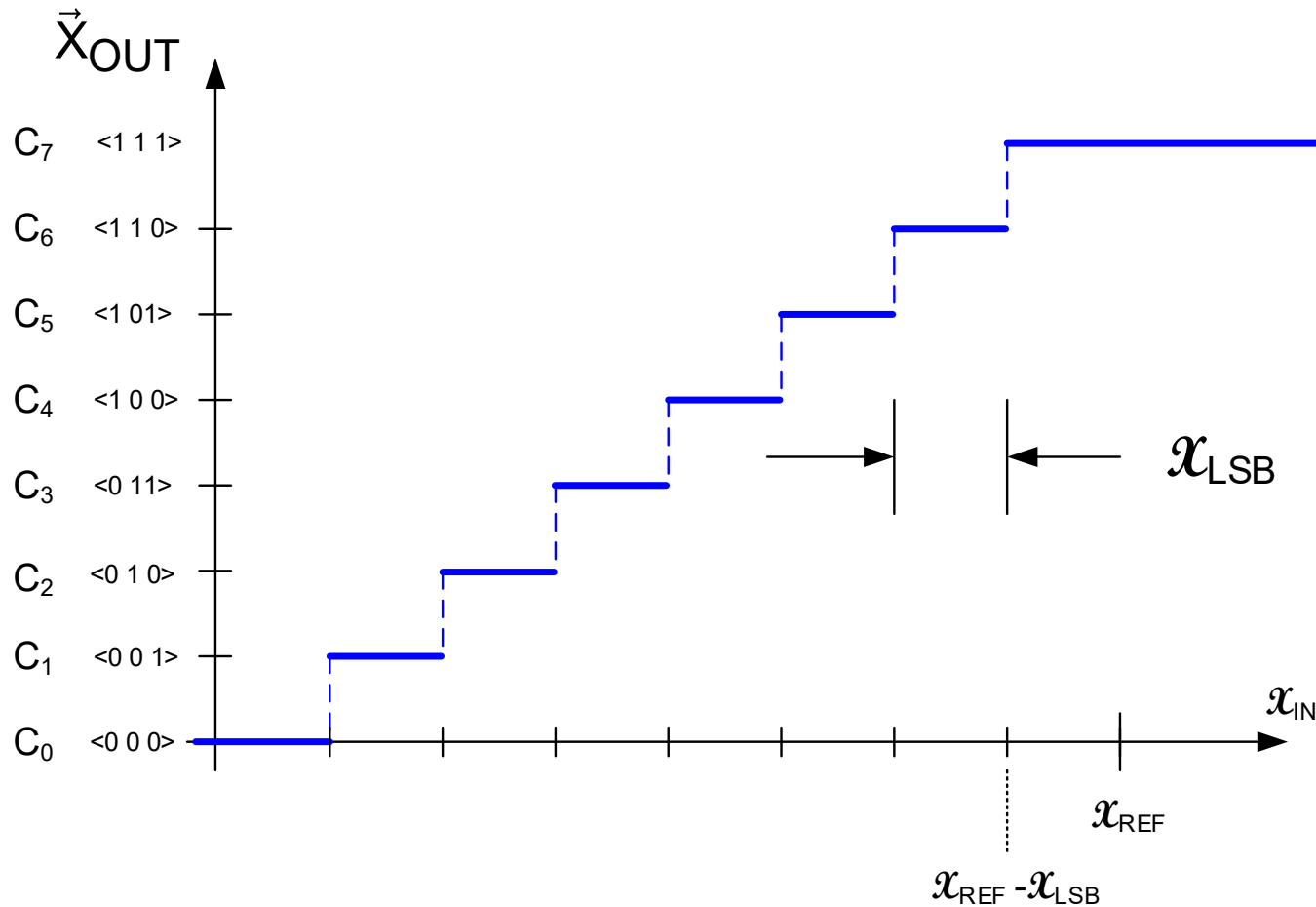
An Ideal ADC is characterized at low frequencies by its static performance

A/D Converters

An Ideal ADC transfer characteristic (3-bits)



$$\vec{X}_{OUT} = \langle d_{n-1}, d_{n-2}, \dots, d_0 \rangle$$



$$x_{LSB} = \frac{x_{REF}}{2^n}$$

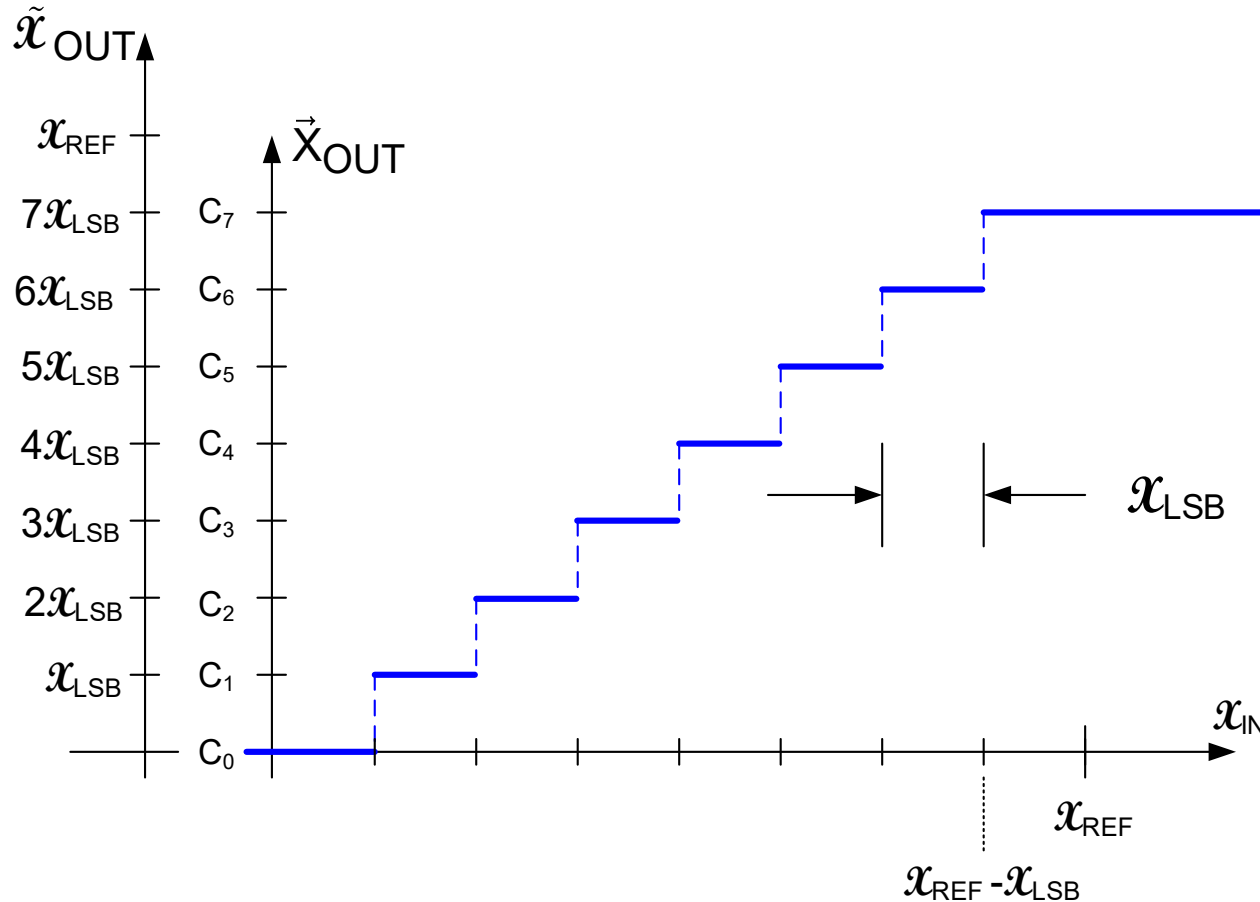
A/D Converters



An Ideal ADC transfer characteristic (3-bits)

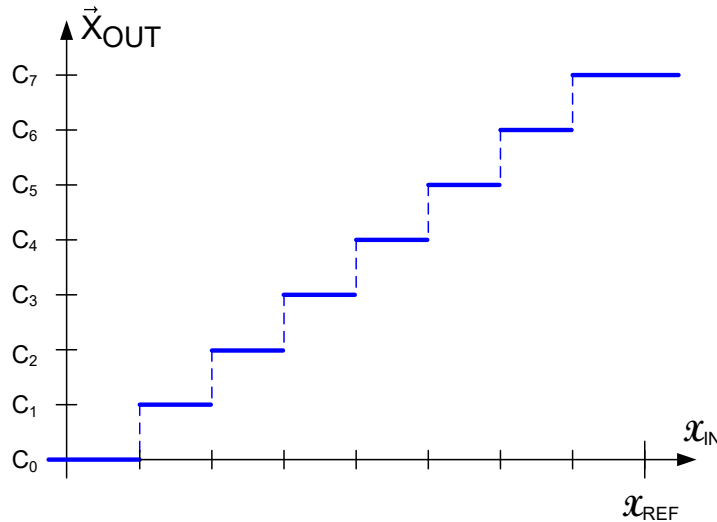
$$\vec{X}_{OUT} = \langle d_{n-1}, d_{n-2}, \dots, d_0 \rangle$$

$$x_{LSB} = \frac{x_{REF}}{2^n}$$



The second vertical axis, labeled \tilde{x}_{OUT} , is the interpreted value of \vec{X}_{OUT}

A/D Converters



For this ideal ADC

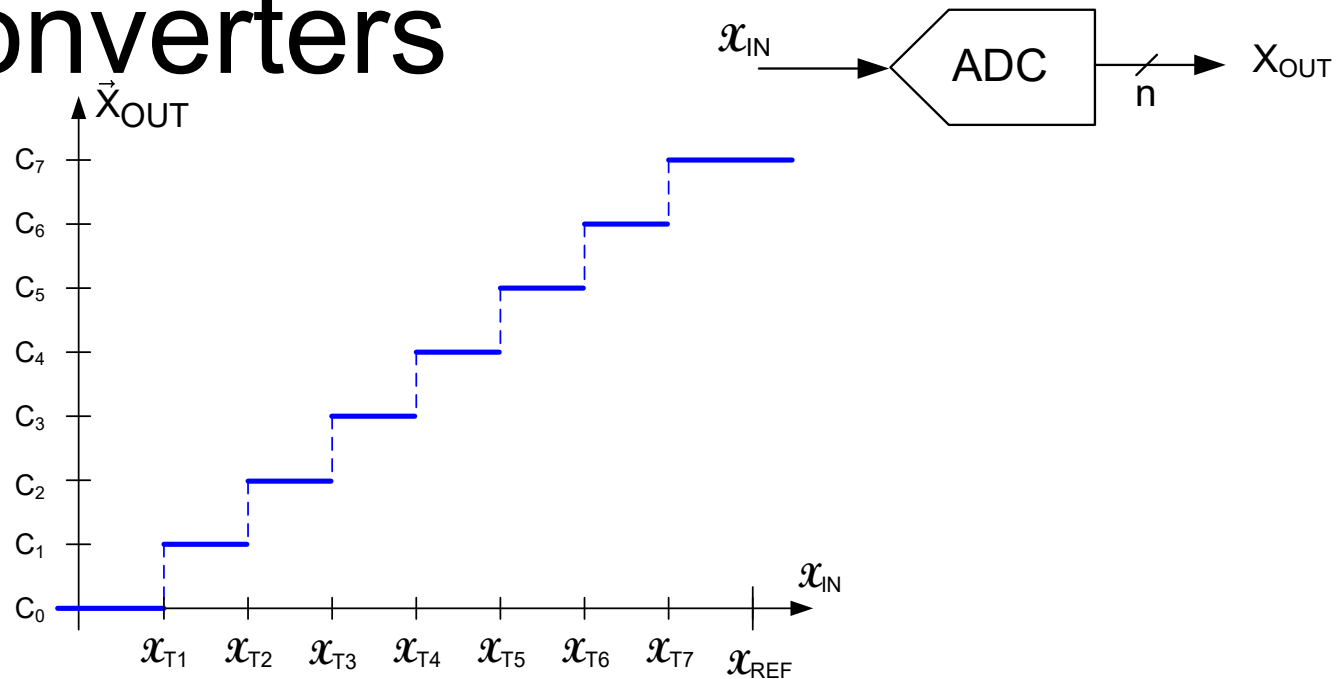
$$x_{\text{REF}} \left(\frac{d_{n-1}}{2} + \frac{d_{n-2}}{4} + \frac{d_{n-3}}{8} + \dots + \frac{d_1}{2^{n-1}} + \frac{d_0}{2^n} \right) = x_{\text{IN}} + \varepsilon$$

where ε is small (typically less than 1LSB)

$$x_{\text{REF}} \sum_{j=1}^n \frac{d_{n-j}}{2^j} = x_{\text{IN}} + \varepsilon$$

- Number of bins gets very large for n large
 - Spacing between break points is $x_{\text{REF}}/2^n$ and gets very small for n large
- ε is the **quantization error** and is inherent in any ADC

A/D Converters



Transition Points

- Actual values of x_{IN} where transitions occur are termed transition points or break points
- For an ideal n-bit ADC, there are $2^n - 1$ transition points
- Ideally the transition points are all separated by 1 LSB -- $X_{LSB} = X_{REF} / 2^n$
- Ideally the transition points are uniformly spaced
- In an actual ADC, the transition points will deviate a little from their ideal location

Labeling Convention:

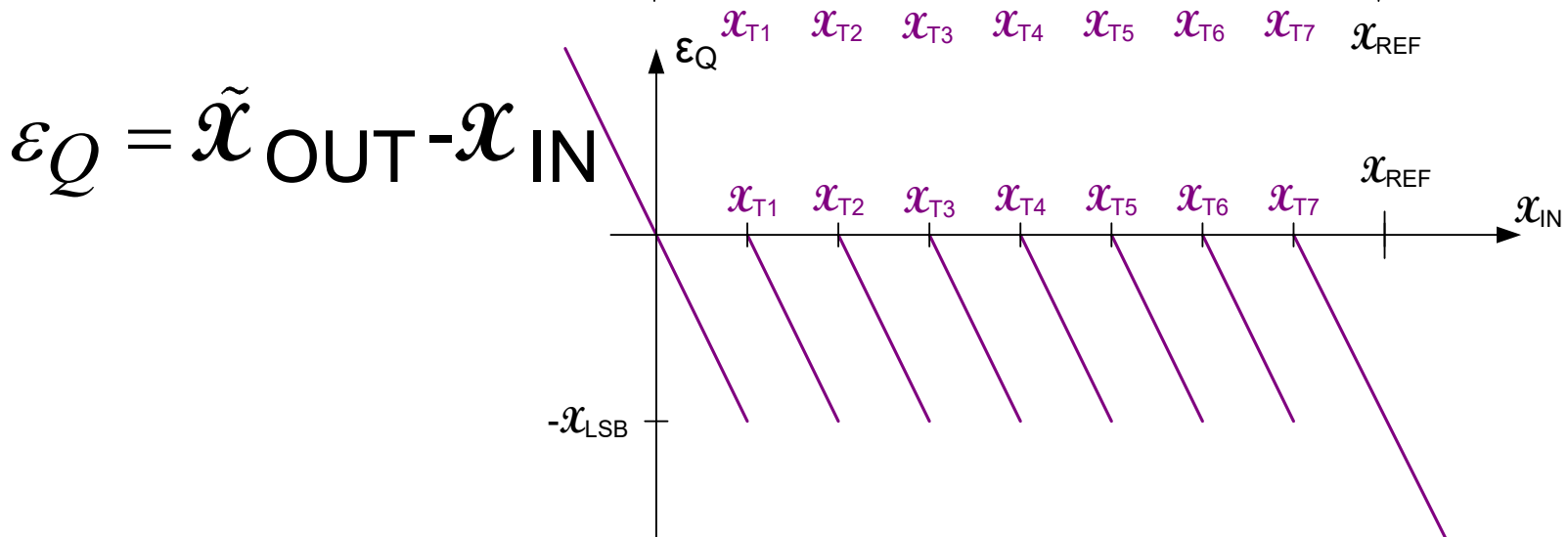
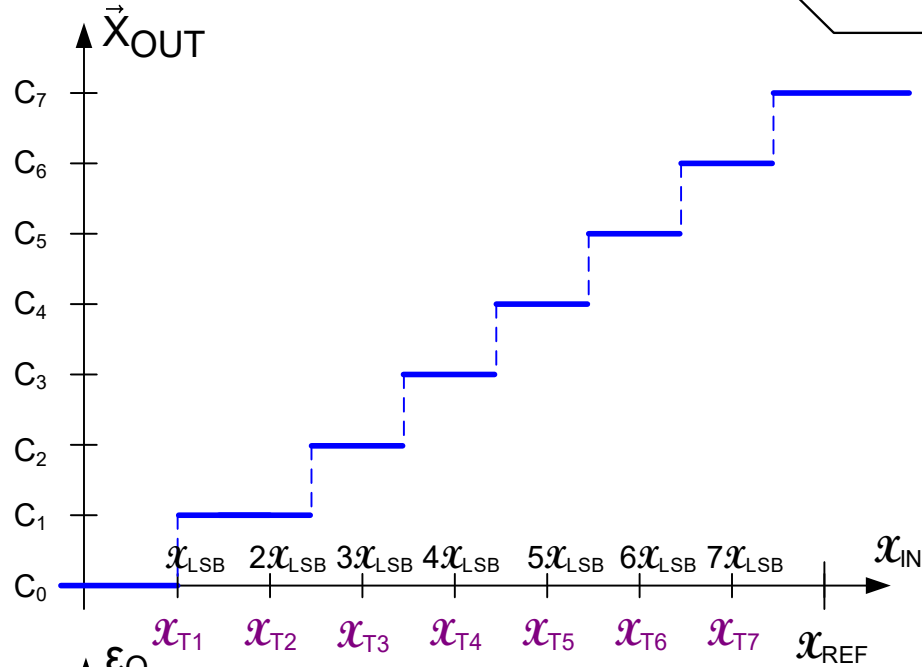
We will define the transition point x_{Tk} to be the break point where the transition in the code output to code C_k occurs. This seemingly obvious ordering of break points becomes ambiguous, though, when more than one break points cause a transition to code C_k which can occur in some nonideal ADCs

A/D Converters



Quantization Errors

$$x_{T1} = x_{LSB}$$



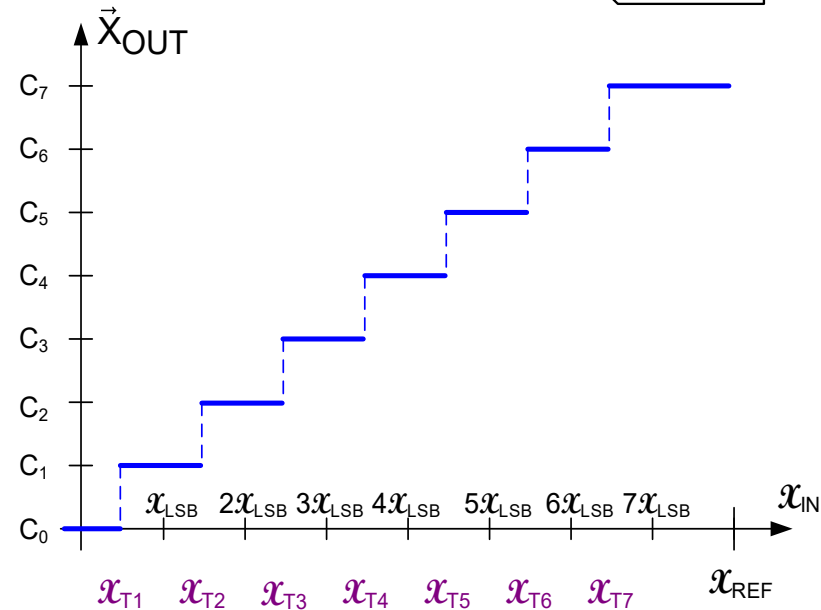
Magnitude of ϵ_Q bounded by x_{LSB} for an ideal A/D

A/D Converters

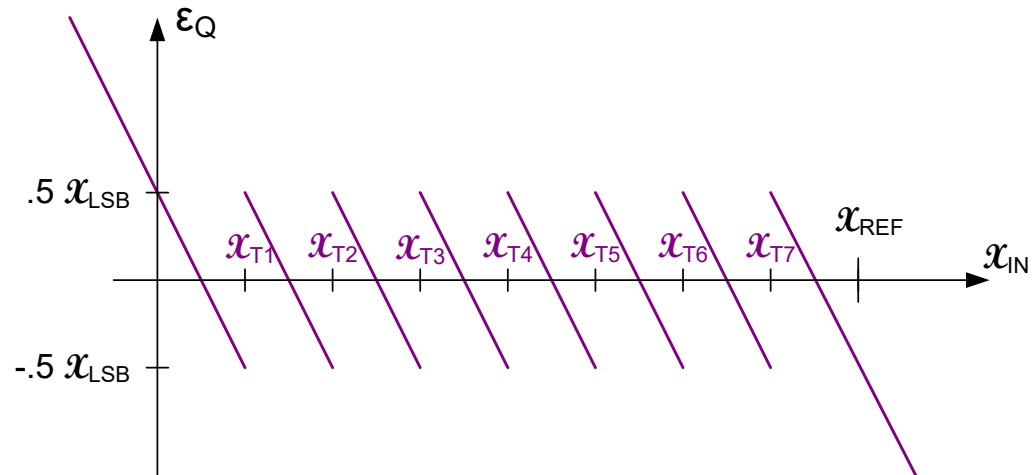
Quantization Errors

Another Ideal ADC

$$x_{T1} = x_{LSB}/2$$



$$\varepsilon_Q = \tilde{x}_{OUT} - x_{IN}$$



Magnitude of ε_Q bounded by $\frac{1}{2} x_{LSB}$

Is the performance of this ideal ADC really better than that of the previous ideal ADC?

Data Converter Architectures



- Large number of different circuits have been proposed for building data converters
- Often a dramatic difference in performance from one structure to another
- Performance of almost all structures are identical if ideal components are used
- Much of data converter design involves identifying the problems associated with a given structure and figuring out ways to reduce the effects of these problems
- Critical that all problems that are significant be identified and solved
- Many of the problems are statistical in nature and implications of not solving problems are in a yield loss that may be dramatic

Data Converter Design Approach

Ideally all performance limitations below performance threshold

Performance Threshold



Data Converter Design Approach

Often one or two nonideal effects above performance threshold



Often try to push dominant nonideal effect down

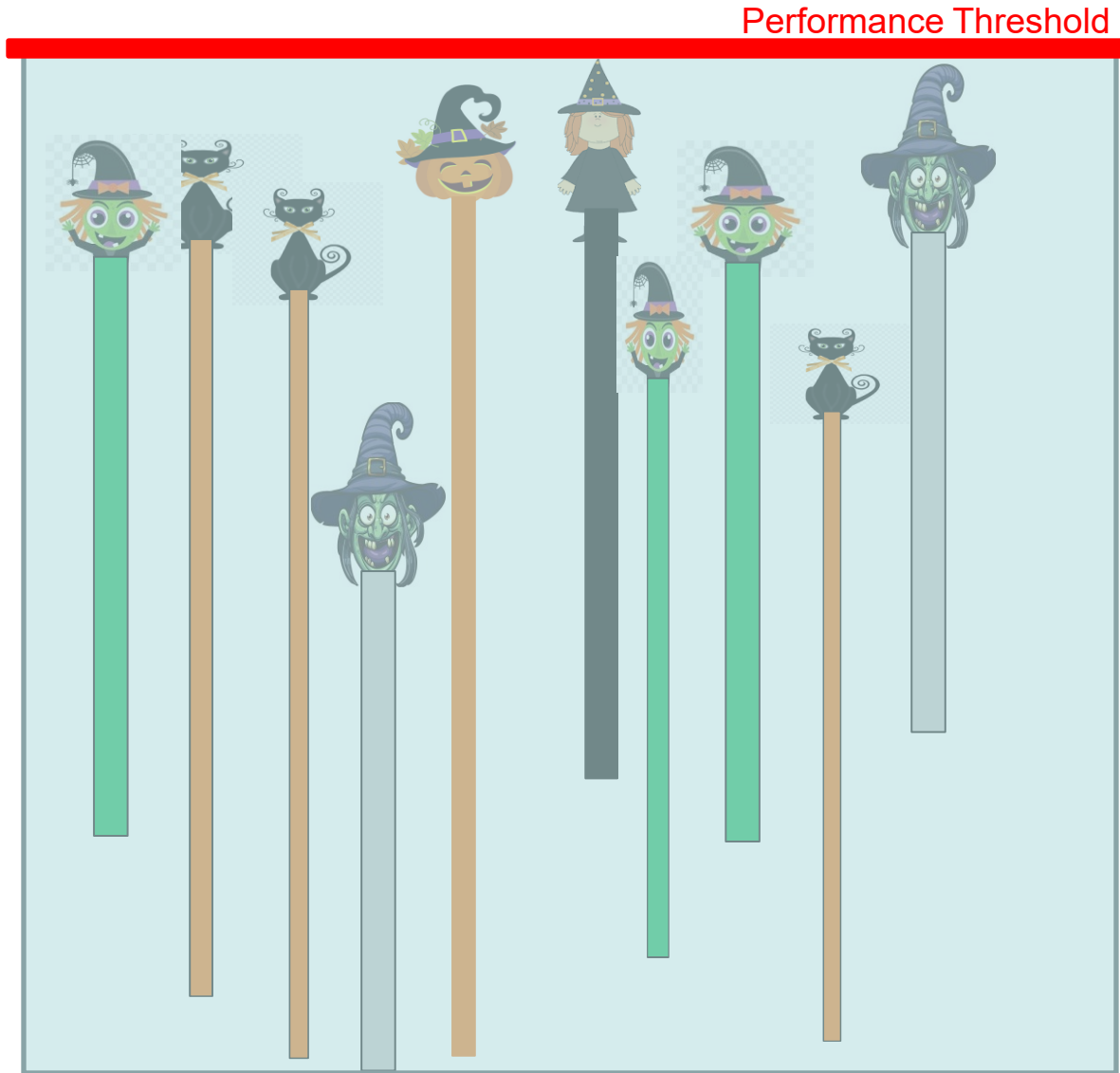
Performance Threshold



Data Converter Design Approach

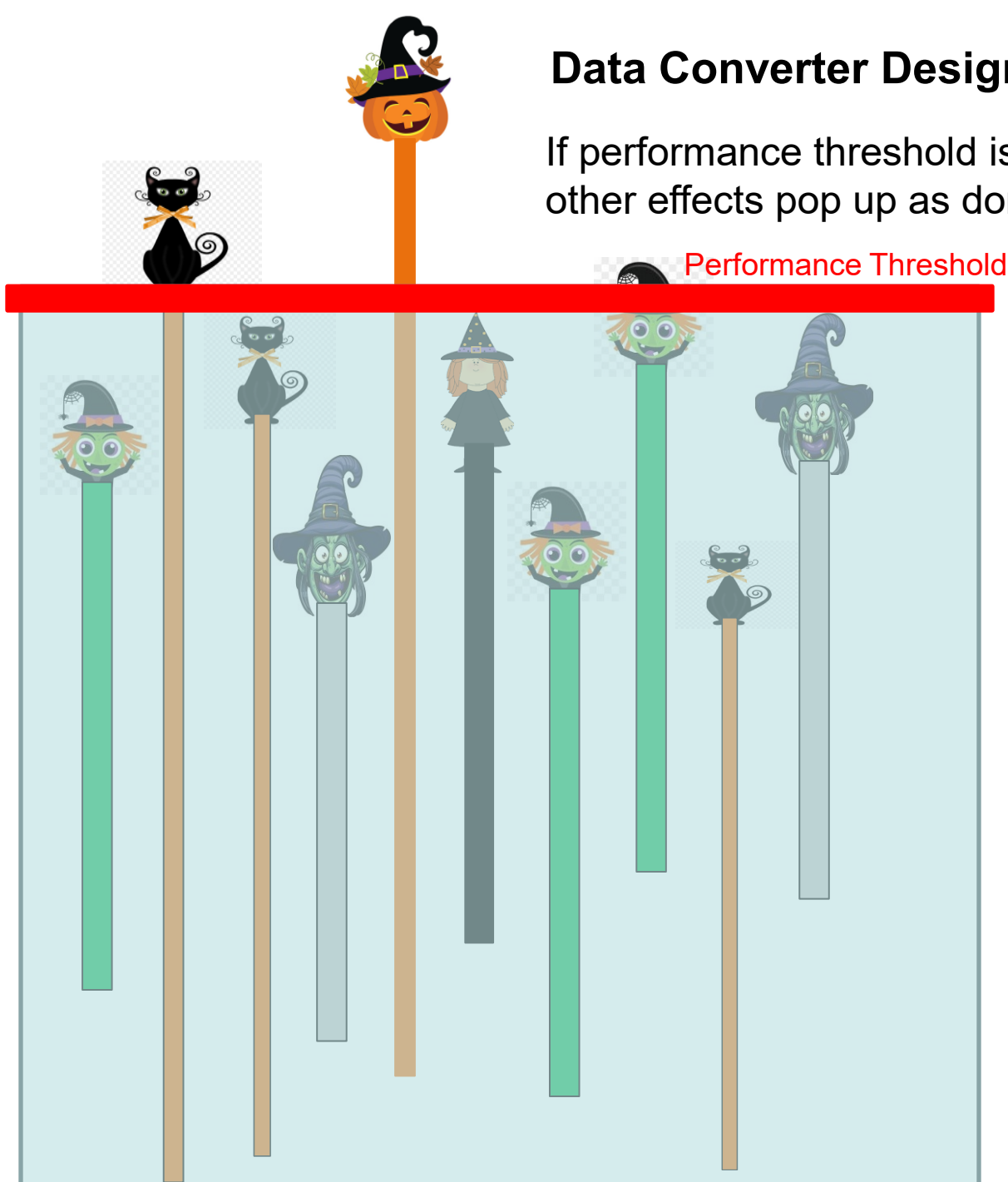
Often one or two nonideal effects above performance threshold

Often try to push dominant nonideal effect down



Data Converter Design Approach

If performance threshold is made “higher”, other effects pop up as dominant



Data Converter Design Approach

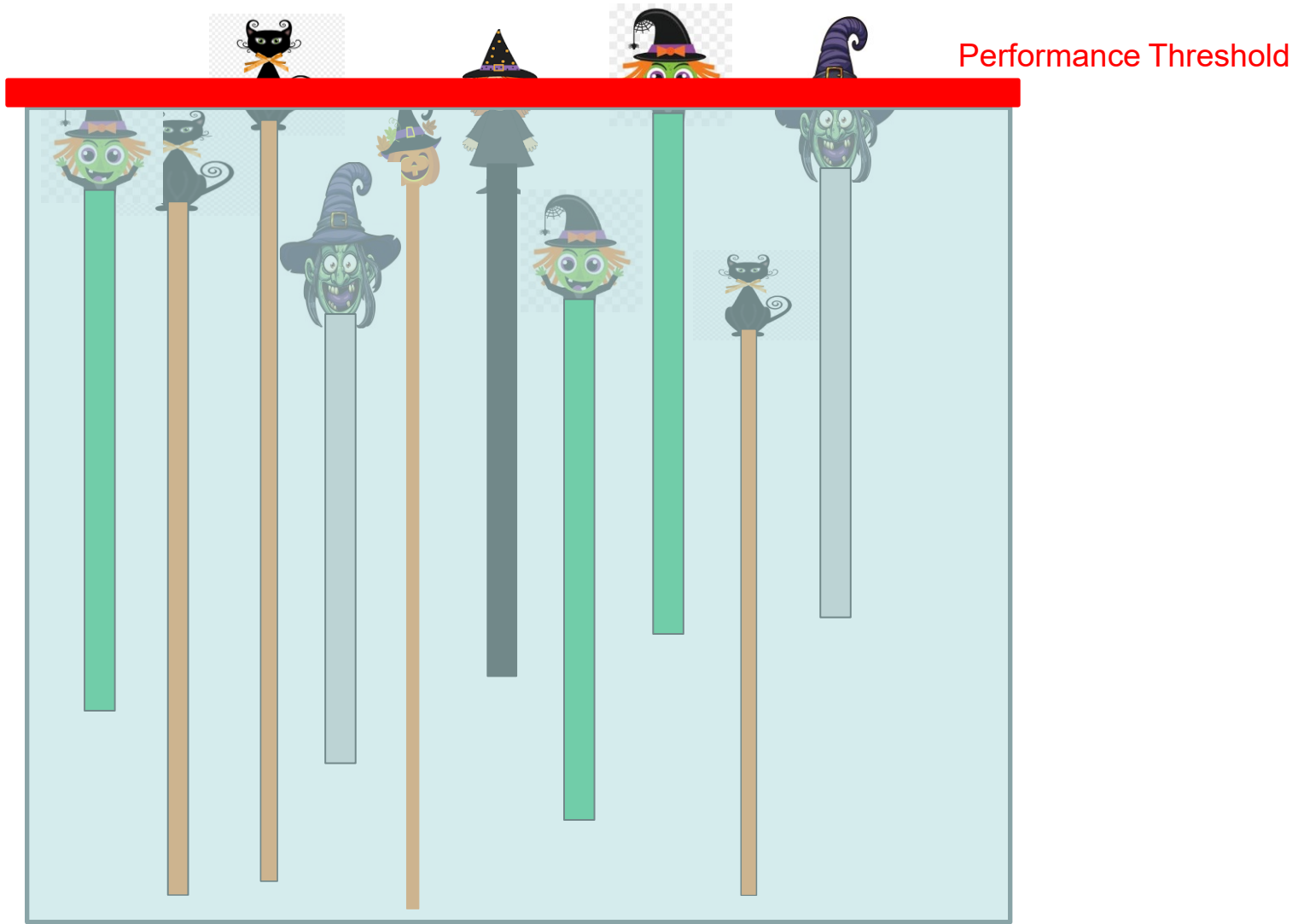
Push them down too

Performance Threshold



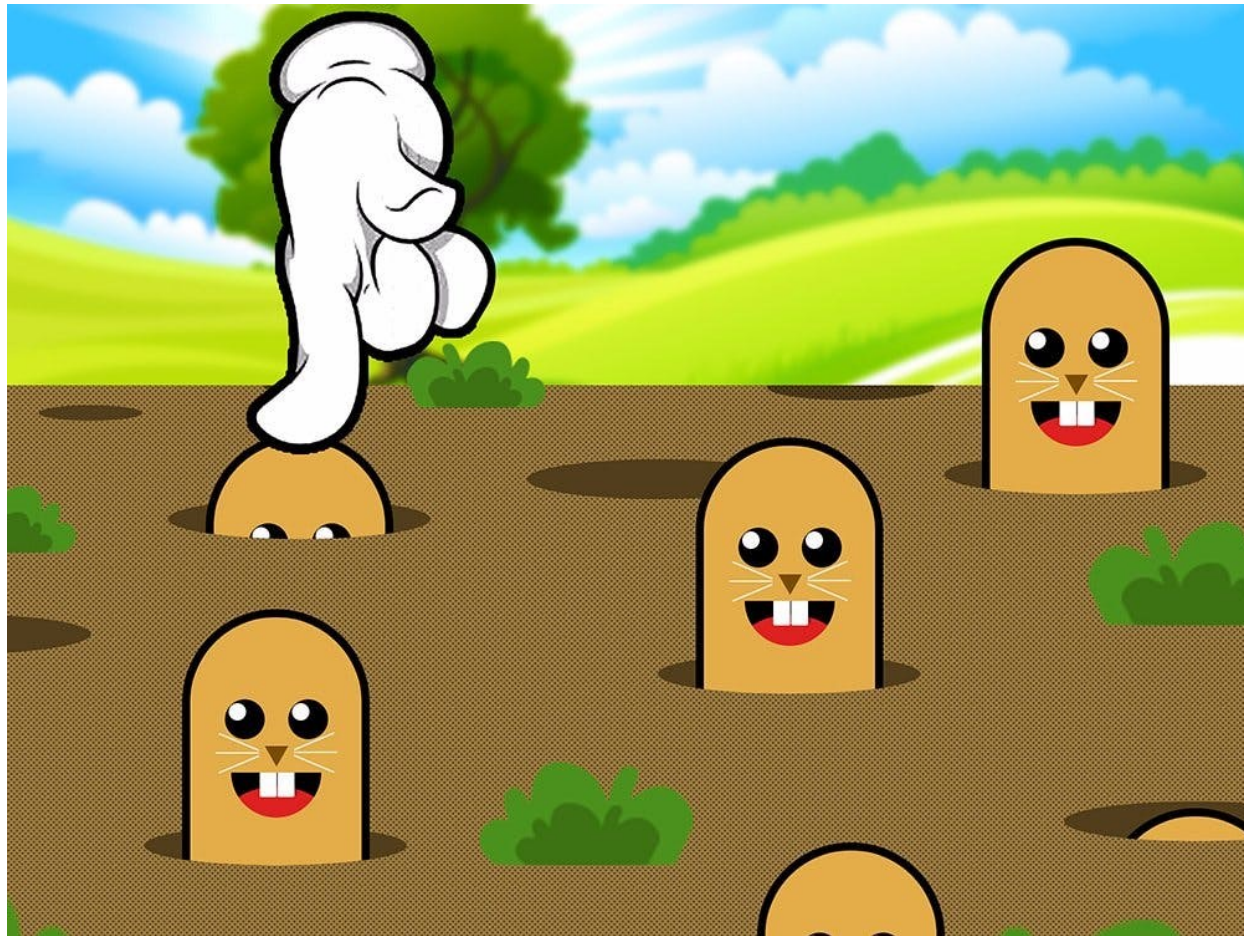
Data Converter Design Approach

Ultimately enhancing performance threshold makes it more difficult to further improve performance



A conceptual view:

Data Converter Design Process Much Like “Whack a Mole” game



Data Converter Architectures



Strategy for discussing data converters

- Briefly look at some different data converter architectures
- Detailed discussion of performance parameters for data converters
- More detailed discussion of data converter architectures

Data Converter Architectures



Nyquist Rate

Flash

Charge Redistribution

Pipeline

Two-step and Multi-Step

Interpolating

Algorithmic/Cyclic

Successive Approximation (Register) SAR

Single Slope / Dual Slope

Subranging

Folded

Interleaved

Current Steering

R-string

Charge Redistribution

Algorithmic

R-2R (ladder)

Pipelined

Subranging

Over-Sampled (Delta-Sigma)

Discrete-time

First-order/Higher Order

Continuous-time

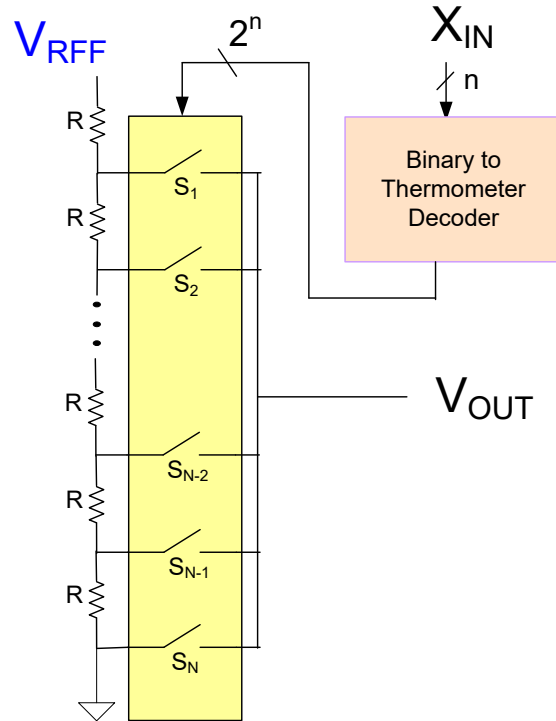
Discrete-time

First-order/Higher Order

Continuous-time

Data Converter Architectures

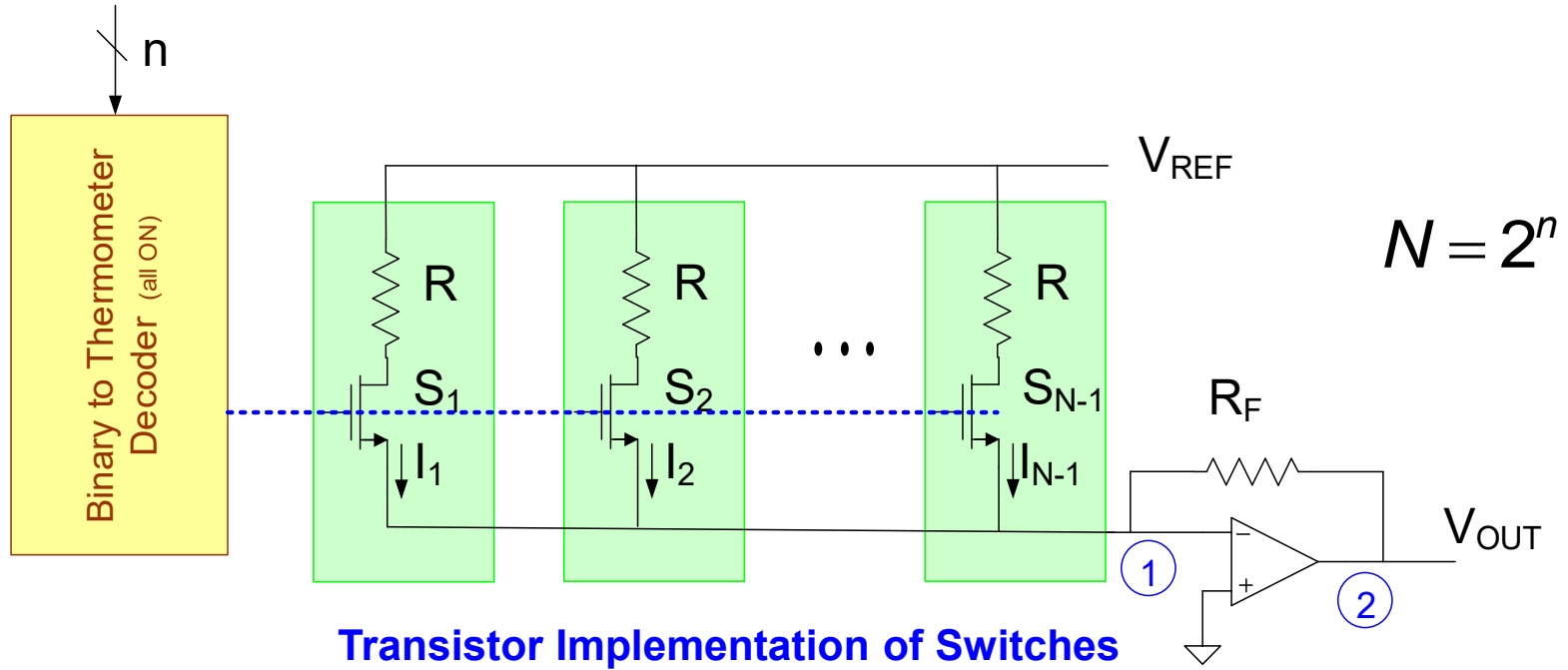
R-String DAC



Basic R-String DAC including Logic to Control Switches

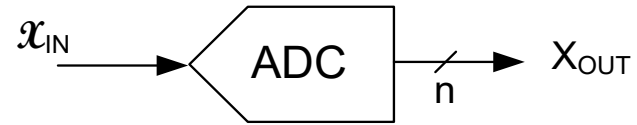
Data Converter Architectures

Current Steering DAC

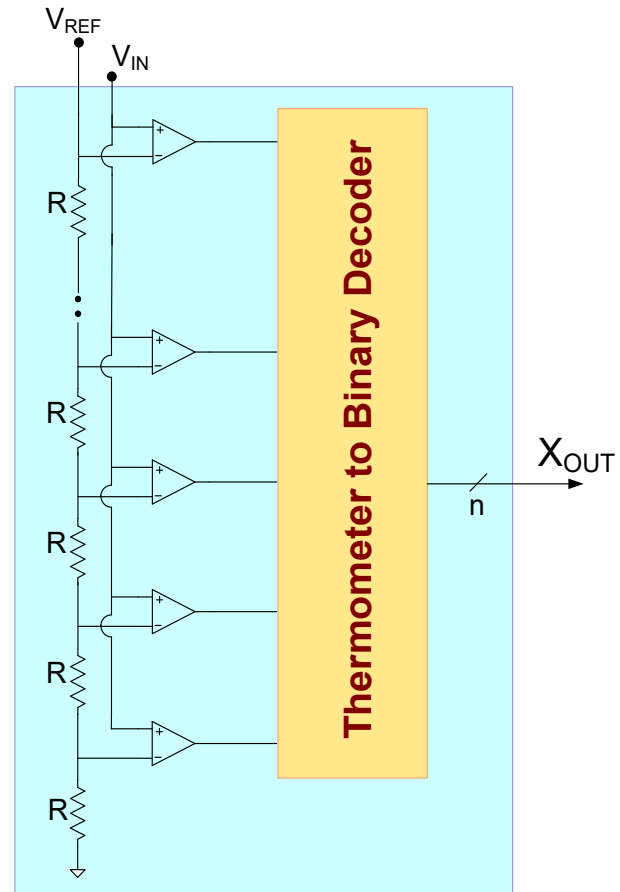


Assume k switches are on $0 < k < N-1$ as determined by digital input code

Data Converter Architectures



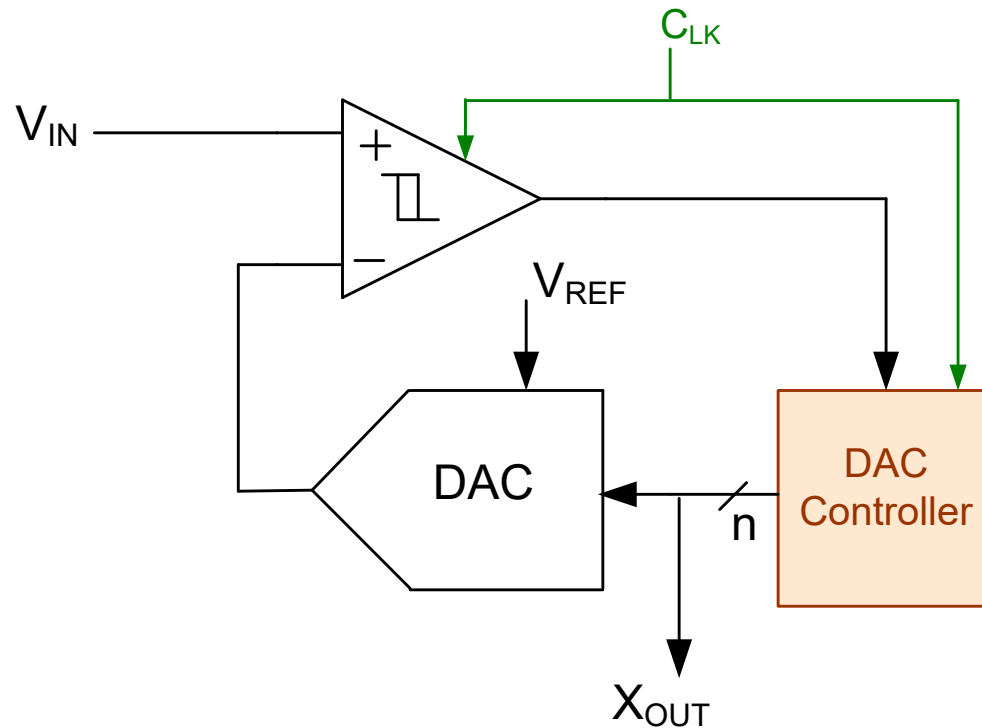
Flash



Data Converter Architectures



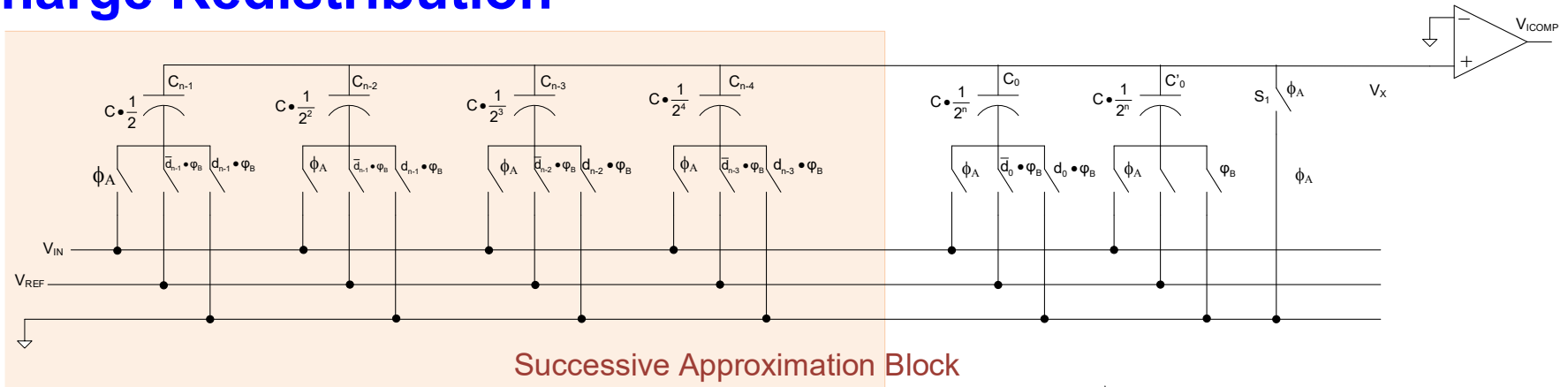
Successive Approximation Register (SAR)



Data Converter Architectures



Charge Redistribution



Successive Approximation Block

Redistribute charge with switches to drive V_X to 0

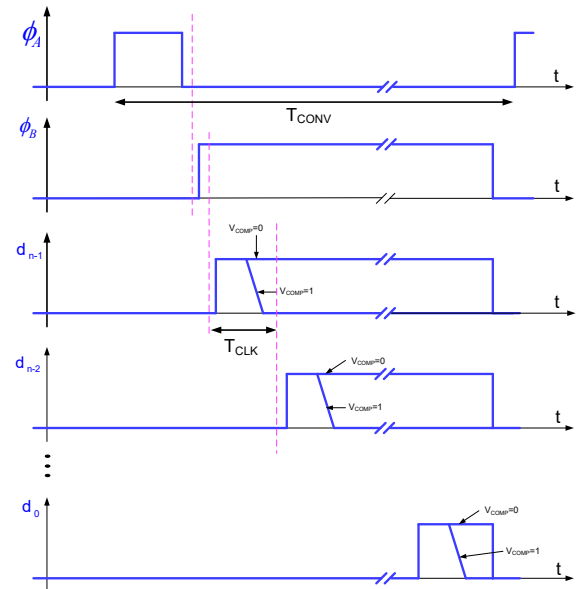
$$Q_{SAM} = V_{IN} \left(\sum_{i=0}^{n-1} C_i + [C'_0] \right) = V_{IN} \left(\sum_{i=0}^{n-1} \frac{C}{2^{n-i}} + \left[\frac{C}{2^n} \right] \right) = V_{IN} C$$

$$Q_{REDIS} = V_{REF} \sum_{i=0}^{n-1} d_i \frac{C}{2^{n-i}}$$

$$Q_{SAM} = Q_{REDIS}$$

$$V_{REF} \sum_{i=0}^{n-1} d_i \frac{C}{2^{n-i}} = V_{IN} C$$

$$V_{IN} = V_{REF} \sum_{i=0}^{n-1} \frac{d_i}{2^{n-i}}$$





Stay Safe and Stay Healthy !

End of Lecture 25